

- Wide Supply Voltage Range: 4.5V - 24V
- 4A Peak Source Current and 4A Peak Sink Current
- Stackable Output for Higher Driving Capability
- Negative Input Voltage Capability: Down to -5V
- TTL Compatible Input Logic Threshold
- Propagation Delay: 13ns
- Typical Rising and Falling Times: 8ns
- Typical Delay Matching: 1ns
- Low Quiescent Current: 55uA
- Output Low When Input Floating
- Independent Enable Logic for Each Channel
- Thermal Shutdown Protection: 170°C
- Available in SOP-8 and eMSOP-8 Package

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Revision 1.0: Released to Production.

Revision 1.3: Update package information, ABS max, add EC table max limit

Revision 1.4: Format adjustment

Revision 1.5: Correct package information in Page 16

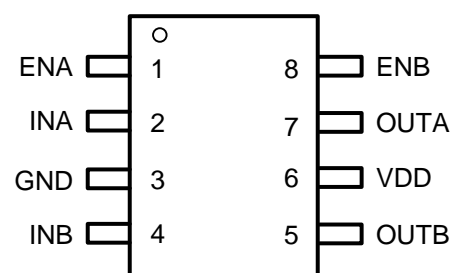
PART NUMBER	PACKAGE MARKING	PACKAGE DISCRIPTION
SCT52240STD	2240	SOP-8
SCT52240MTE	2240	eMSOP-8

1) For Tape & Reel, Add Suffix R (e.g. SCT52240STDR).

Over operating free-air temperature unless otherwise noted⁽¹⁾

DESCRIPTION	MIN	MAX	UNIT
ENA, ENB	-0.3	26	V
INA, INB	-5	26	V
OUTA, OUTB	-0.3	VDD+0.3	V
OUTA, OUTB (Pulse<0.2us)	-3	VDD+3	V
VDD	-0.3	26	V
Operating junction temperature T _J ⁽²⁾	-40	150	°C
Storage temperature T _{STG}	-65	150	°C

Top View: SOP-8pin
Plastic



- (1) Stresses beyond those listed under Absolut Maximum Rating may cause device permanent damage. The device is not guaranteed to function outside of its Recommended Operation Conditions.
- (2) The IC includes over temperature protection to protect the device during overload conditions. Junction temperature will exceed 150°C when over temperature protection is active. Continuous operation above the specified maximum operating junction temperature will reduce lifetime

NAME	NO.	PIN FUNCTION
ENA	1	Channel A enable logic input, TTL compatible. Floating logic high.
INA	2	Channel A logic input, TTL compatible. Floating logic low.
GND	3	Power ground. Must be soldered directly to ground plane for thermal performance improvement and electrical contact.
INB	4	Channel B logic input, TTL compatible. Floating logic low.
OUTB	5	Channel B gate driver output
VDD	6	Power Supply, must be locally bypassed by the ceramic cap.

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OUTA	7	Channel A gate driver output
ENB	8	Channel B enable logic input, TTL compatible. Floating logic high.

Over operating free-air temperature range unless otherwise noted

PARAMETER	DEFINITION	MIN	MAX	UNIT
V _{DD}	Supply voltage range	4.5	24	V
V _{INA,INB}	Input voltage range	-5	24	
T _J	Operating junction temperature	-40	150	°C

PARAMETER	DEFINITION	MIN	MAX	UNIT
V _{ESD}	Human Body Model (HBM), per ANSI-JEDEC-JS-001-2014 specification, all pins ⁽¹⁾	-2	+2	kV
	Device Model (CDM), per ANSI-JEDEC-JS-002-2014 specification, all pins ⁽¹⁾	-0.5	+0.5	kV

(1) HBM and CDM stressing are done in accordance with the ANSI/ESDA/JEDEC JS-001-2014 specification

PARAMETER	THERMAL METRIC	SOP-8L	eMSOP-8	UNIT
R	Junction to ambient thermal resistance ⁽¹⁾	130	72	°C/W
R	Junction to case thermal resistance ⁽¹⁾			

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V_{DD}=12V, T_J=-40°C~150°C, typical values are tested under 25°C.

SYMBOL	PARAMETER	TEST CONDITION	MIN	TYP	MAX	UNIT
Power Supply and Output						
	Operating supply voltage		4.5		24	V
D_UVLO	Input UVLO Hysteresis	V _{DD} rising		4.2 300	4.5	V mV
	Supply current	EN=V _{DD} =3.5V, INA=INB=GND		55	115	uA
		EN=V _{DD} =12V, INA=INB=GND		120	190	uA

INPUTS

INA,INB_H	Input logic high threshold			2.1	2.4	V
INA,INB_L	Input logic low threshold		0.8	1		V
IN_Hys	Hysteresis			1.1		V
INA,ENB_H	Enable logic high threshold			2.1	2.4	V
INA,ENB_L	Enable logic low threshold		0.8	1		V
	Hysteresis			1.1		V

OUTPUTS

SCT52240

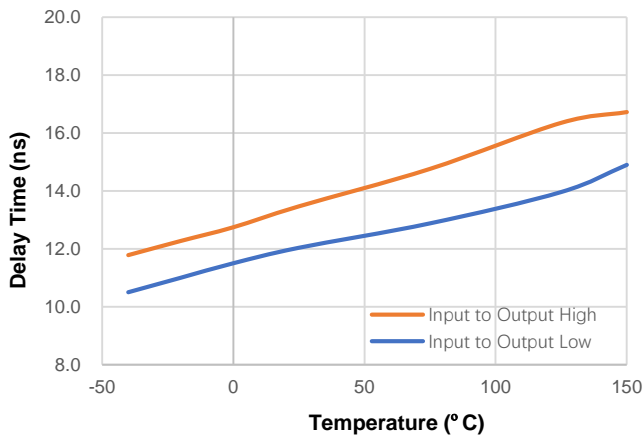


Figure 7. Input to Output Propagation Delay vs Temperature

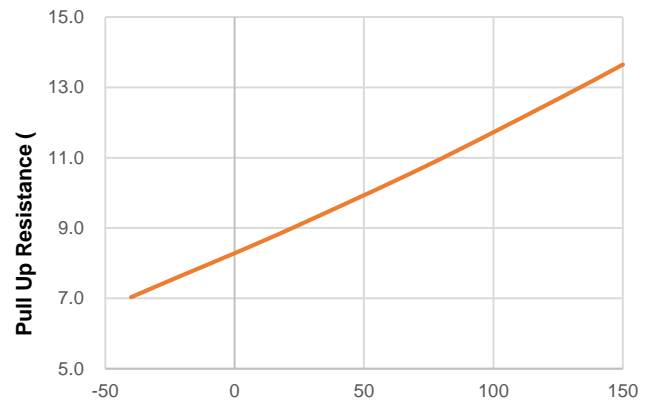
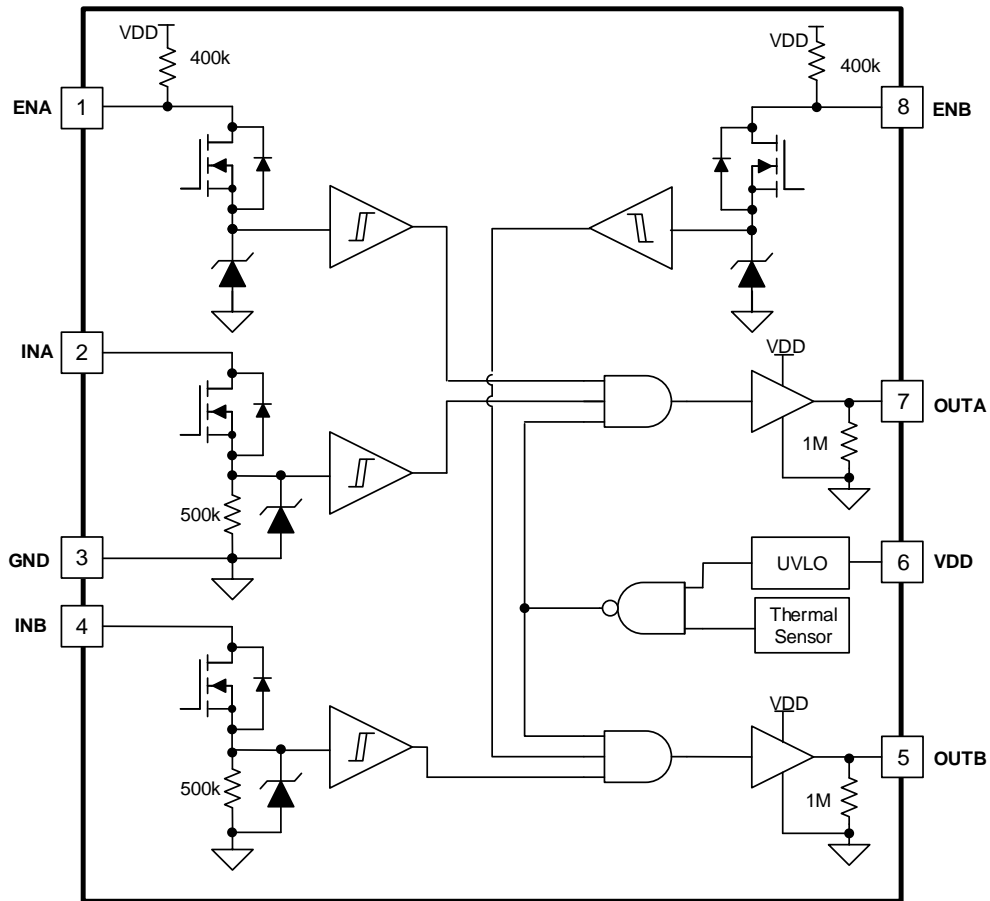


Figure 8. ROH vs Temperature

Figure 9. ROL vs Temperature

Figure 10. Operation Supply Current vs Frequency, $C_{OUT}=1nF$



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Overview

The SCT52240 is a dual-channel non-invertible high-speed low side driver with supporting

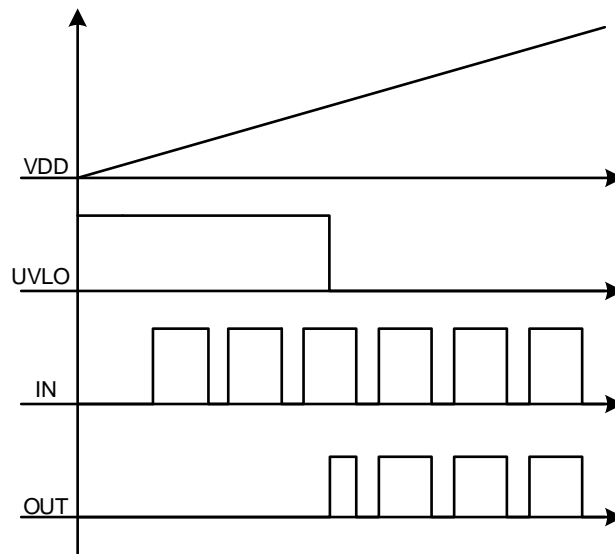


Figure 11. SCT52240 Output Vs VDD

Enable Function

SCT52240 provides independent enable pins ENA and ENB for external control of each channel operation. The enable pins are based on a TTL compatible input-threshold logic that is independent of the supply voltage and is effectively controlled with logic signals from 3.3-V and 5-V microcontrollers. When applying a voltage higher than the high threshold (typical 2.1V) the pin, the SCT52240 enables all functions and starts gate driver operation. Driver operation is disabled when ENx voltage falls below its lower threshold (typical 1V). The ENx pins are internally pulled up to VDD with 400k pullup resistors. Hence, the ENx pins are left floating or Not Connected (N/C) for standard operation, where the enable feature is not required.

Input Stage

The input of SCT52240 is compatible on TTL input-threshold logic that is independent of the VDD supply voltage. With typically high threshold = 2.1 V and typically low threshold = 1 V, the logic level thresholds are conveniently driven with PWM control signals derived from 3.3-V and 5-V digital power-controller devices. Wider hysteresis offers enhanced noise immunity compared to traditional TTL logic implementations, where the hysteresis is typically less than 0.5V. SCT52240 also features tight control of the input pin threshold voltage that ensures stable operation across temperature. The very low input parasitic capacitance on the input pins increases switching speed and reduces the propagation delay.

Output Stage

The SCT52240 output stage features the pull up structure with P-type MOSFET PM1 and N-type MOSFET NM1 in parallel, as shown in Figure 12. PM1 provides the pull up capability when OUT approaches VDD and the NM1 holds off state, which guarantees the driver output is up to VDD rail. The measurable on-resistance R_{OH} in steady state is the conduction resistance of PM1. NM1 provides a narrow instant peak sourcing current up to 4A to eliminate the turn on time and delay. During the output turn on transition, the equivalent hybrid pull on transient resistance is $1.5R_{OL}$, which is much lower than the DC measured R_{OH} .

The N-type MOSFET NM2 composes the output stage pull down structure; the R_{OL} is the DC measurement and represents the pull down impedance. The output stage of SCT52240 provides rail-to-rail operation, and is able to supply 4A sourcing and 4A sinking peak current. The presence of the MOSFET-body diodes also offers low impedance to switching overshoots and undershoots. The outputs of the dual channel drivers are designed to withstand 500-mA reverse current without either damaging the device or logic malfunction.

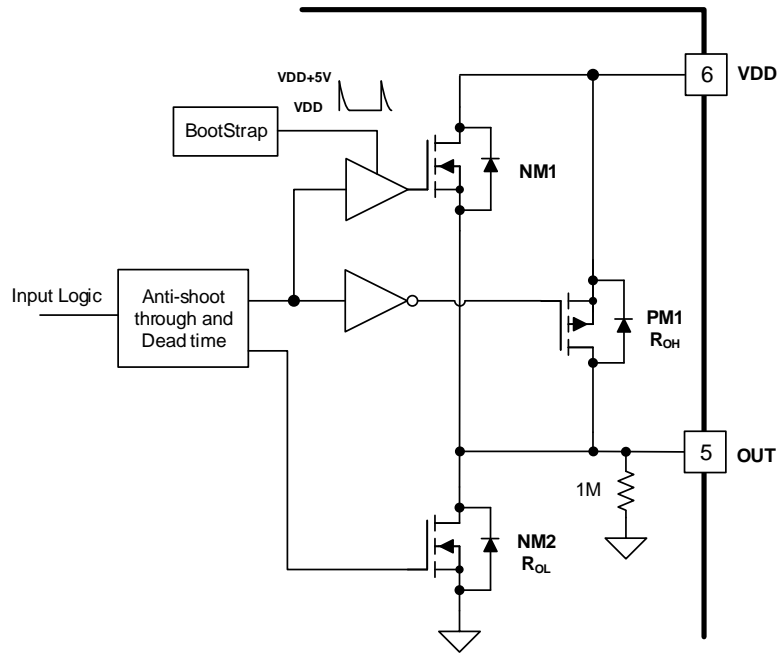


Figure 12. SCT52240 Output Stage

Stackable Output

The SCT52240 features 1ns (typical) delay matching between dual channels, which enables dual channel outputs be stackable when the driven power device required higher driving capability. For example, in a Boost Power Factor Correction converter, there are 2 power MOSFET in parallel to support higher power output capability. The two power MOSFET are preferred to be driven by a common gate control signal. By using SCT52240, the OUTA and OUTB can be connected together to provide the higher driving capability, so does the INA and INB. As a result, a single input signal controls the stacked output combination. To support the stackable output, each channel output stage artificially implements up to 5ns dead-time to avoid the possible shoot-through between two channels as shown Figure 13.

Due to the rising and falling threshold mismatch between INA and INB, cautions must be taken when implementing stackable output of OUTA and OUTB together. The maximum mismatch between INA and INB input threshold is up to 10mV (maximum cross temperature), as a result the allowed minimum slew rate of input logic signal is 2V/us. The following suggestions are recommended when INA and INB connected together and along with the OUTA and OUTB:

1. Apply the fast slew rate dv/dt on input (2 V/us or greater) to avoid the possible shoot-through between OUTA and OUTB output stage.
2. INA and INB must be connected as close to the pins as possible.

Typical Application

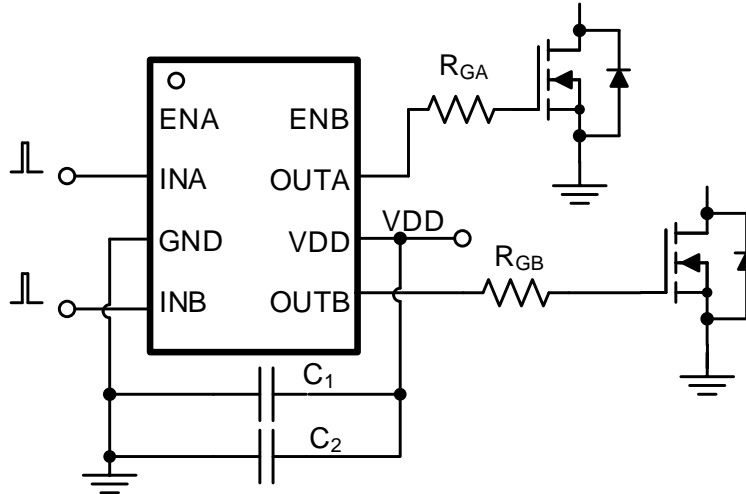


Figure 16. Dual Channel Driver Typical Application

Driver Power Dissipation

Generally, the power dissipated in the SCT52240 depends on the gate charge required of the power device (Q_g), switching frequency, and use of external gate resistors. The SCT52240 features very low quiescent currents and internal logic to eliminate any shoot-through in the output driver stage, their effect on the power dissipation within the gate driver is negligible.

For the pure capacitive load, the power loss of each channel in SCT52240 is:

(1)

Where

- V_{DD} is supply voltage
- C_{Load} is the output capacitance
- F_{SW} is the switching frequency

For the the switching load of power MOSFET, the power loss of each channel in the SCT52240 is shown in equation (2), where charging a capacitor is determined by using the equivalence $Q_g = C_{LOAD}V_{DD}$. The gate charge includes the effects of the input capacitance plus the added charge needed to swing the drain voltage of the power device as it switches between the ON and OFF states. Manufacturers provide specifications that provide the typical and maximum gate charge, in nC, to switch the device under specified conditions.

(2)

Where

- Q_g is the gate charge of the power device
- f_{sw} is the switching frequency
- V_{DD} is the supply voltage

If R_G applied between driver and gate of power device to slow down the power device transition, the power dissipation of the driver shows as below:

(3)

Where

- R_{OH} is the equivalent pull up resistance of SCT52240
- R_{OL} is the pull down resistance of SCT52240
- R_G is the gate resistance between driver output and gate of power device.

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Application Waveforms

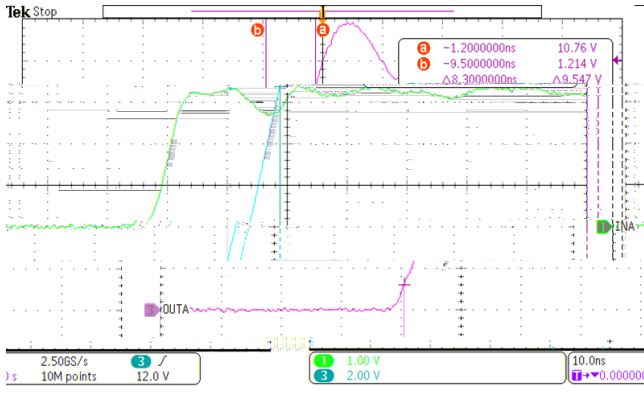


Figure 17. Driver Switching ON

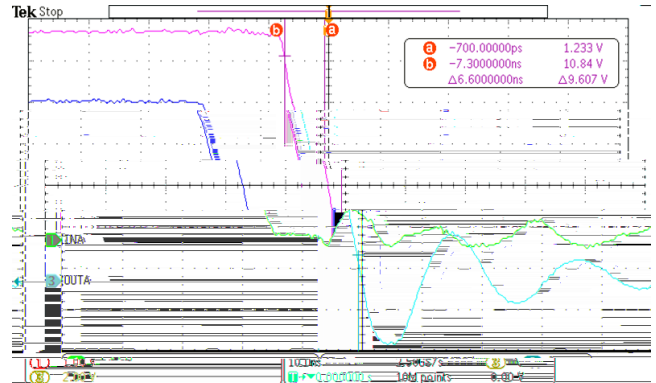


Figure 18. Driver Switching OFF

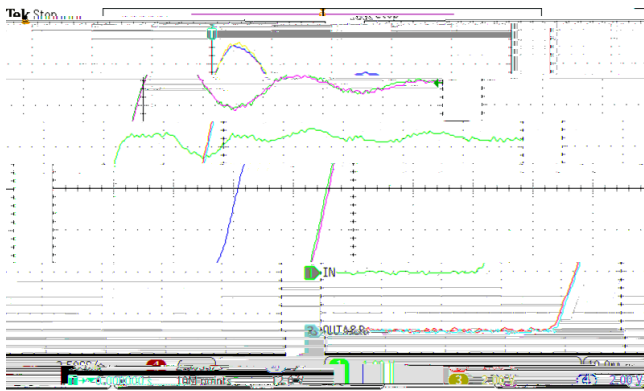


Figure 19. Delay Matching Rise

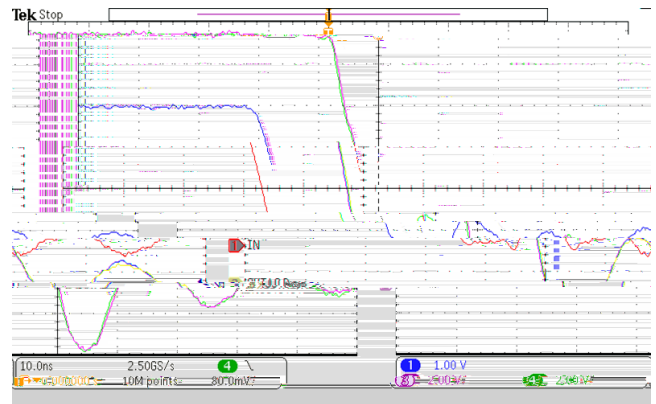


Figure 20. Delay Matching Fall

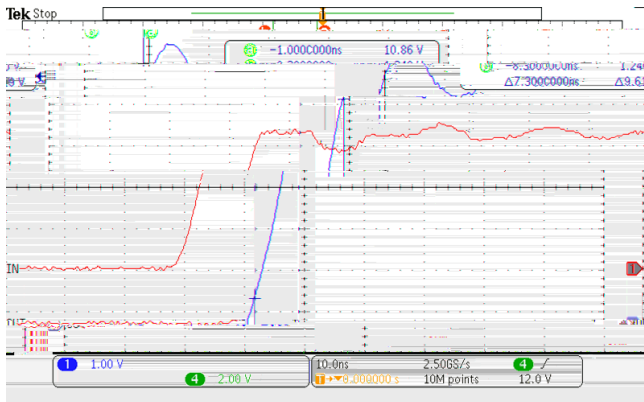


Figure 21. Stackable Output Rise

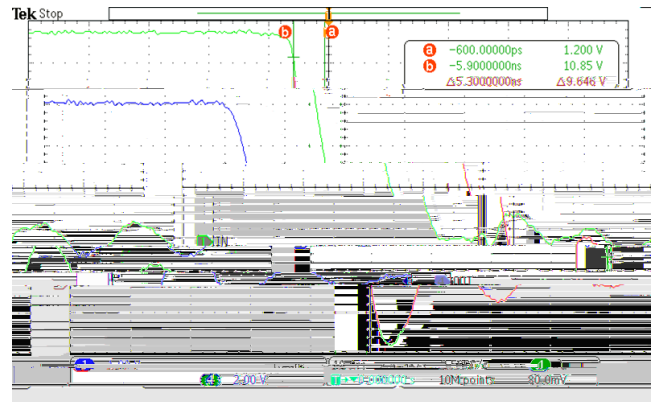
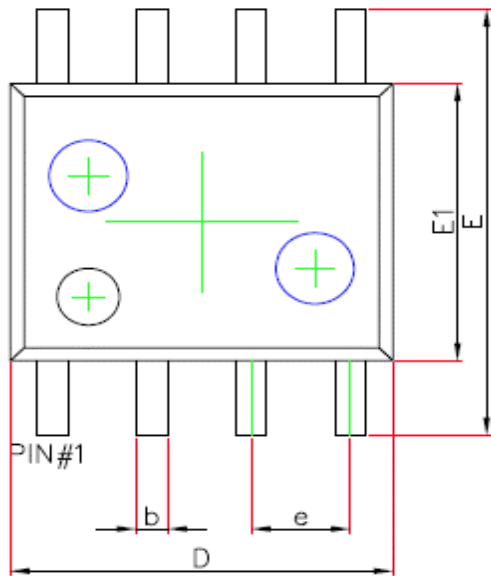
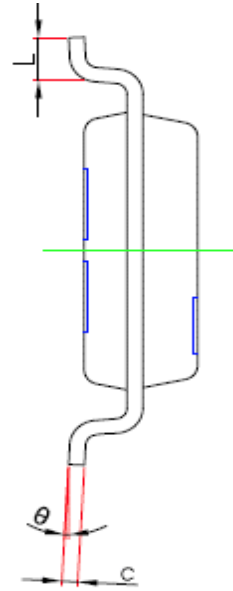


Figure 22. Stackable Output Fall

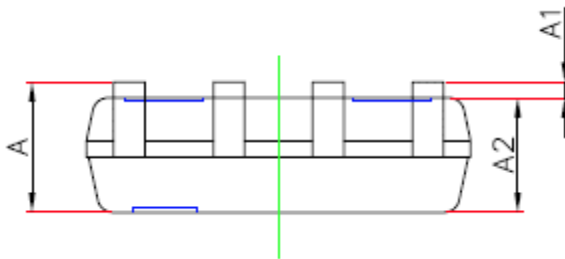
SCT52240



TOP VIEW



BOTTOM VIEW

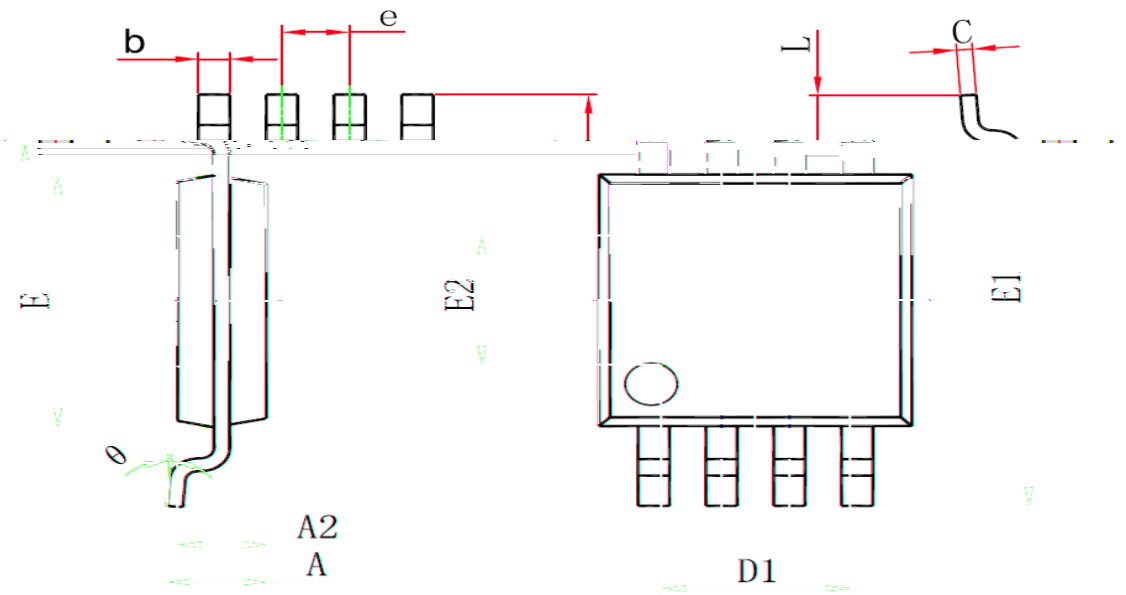


SIDE VIEW

NOTE:

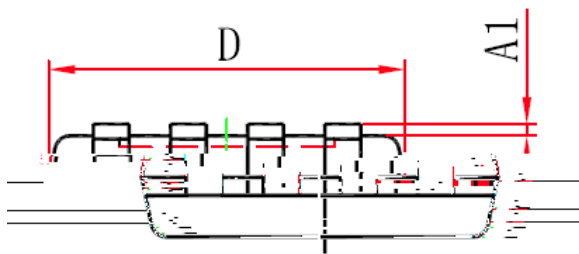
1. Drawing proposed to be made a JEDEC package outline MO-220 variation.
2. Drawing not to scale.
3. All linear dimensions are in millimeters.
4. Thermal pad shall be soldered on the board.
5. Dimensions of exposed pad on bottom of package do not include mold flash.
6. Contact PCB board fabrication for minimum solder mask web tolerances between the pins.

SYMBOL	Unit: Millimeter		
	MIN	TYP	MAX
A	1.45	---	1.75
A1	0.1	---	0.25
A2	1.35	---	1.55
b	0.33	---	0.51
c	0.17	---	0.25
D	4.7		5.1
E	5.8		6.2
E1	3.8		4.0
e	1.27BSC		
L	0.4		1.27
	0°		8°



TOP VIEW

BOTTOM VIEW



SIDE VIEW

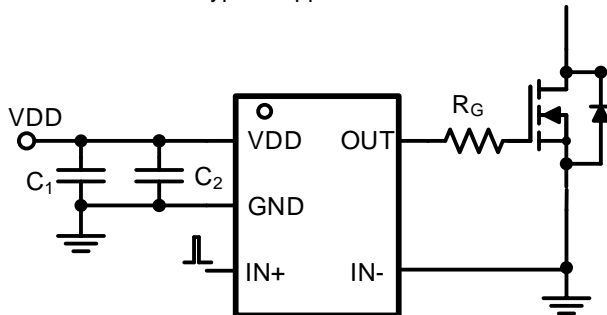
SYMBOL	Unit: Millimeter		
	MIN	TYP	MAX
A	0.82	---	1.1
A1	0.02	---	0.15
A2	0.75	---	0.95
b	0.25	---	0.38
c	0.09	---	0.23
D	2.9		3.1
D1	1.7		1.9
E	2.9		3.1
E1	4.75		5.05
E2	1.45		1.65
e	0.65BSC		
L	0.4		0.8
	0°		6°

NOTE:

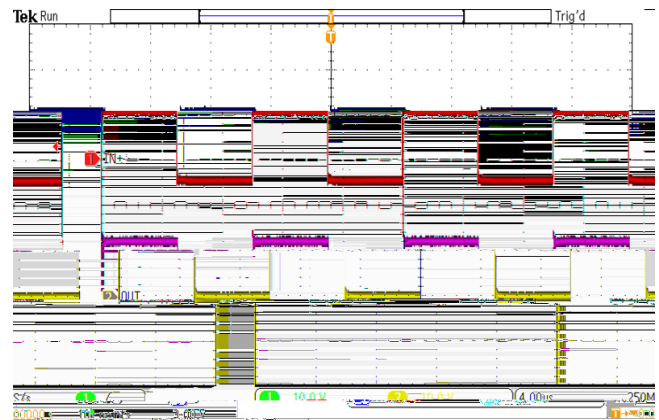
7. Drawing proposed to be made a JEDEC package outline MO-220 variation.
8. Drawing not to scale.
9. All linear dimensions are in millimeters.
10. Thermal pad shall be soldered on the board.
11. Dimensions of exposed pad on bottom of package do not include mold flash.
12. Contact PCB board fabrication for minimum solder mask web tolerances between the pins.

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Single Channel, Non-Inverting MOSFET Gate Drive
Typical Application



Typical Application Waveform



PART NUMBERS	DESCRIPTION	COMMENTS
SCT51240	Up to 24V Supply, 4-A Single Channel High Speed Low Side Driver	<ul style="list-style-type: none"> Compatible for both Inverting and Non-inverting application Supporting down to -5V input

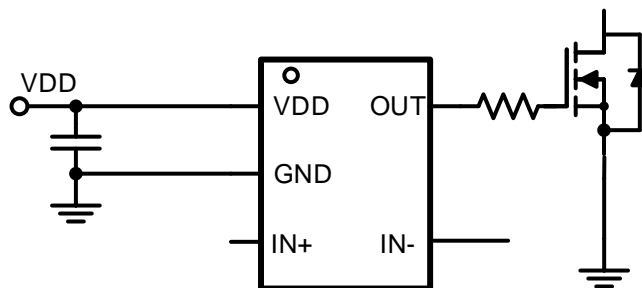


Figure 24. SCT51240 Inverting MOSFET Gate Drive Typical Application

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